CQB100W Series
Application Note V13

ISOLATED DC-DC CONVERTER
CQB100W SERIES
APPLICATION NOTE

Approved By:

Department	Approved By	Checked By	Written By
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Contents

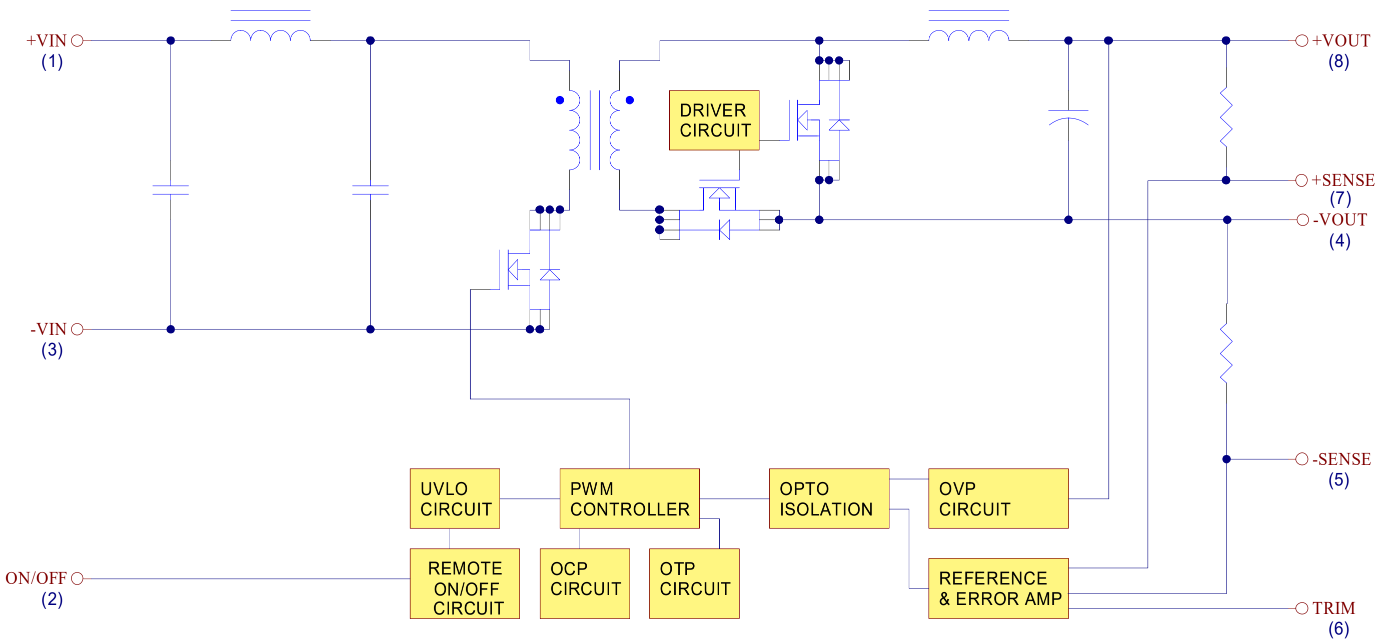
1. Introduction

This specification describes the features and functions of Cincon's CQB100W series of isolated DC-DC Converters. These are highly efficient, reliable and compact, high power density, single output DC/DC converters. The modules can be used in the field of telecommunications, data communications, wireless communications, servers etc. The CQB100W series can deliver up to 30A output current and provide a precisely regulated output voltage over a wide range of input voltages (Vi=9-36Vdc or Vi=18-75Vdc). The modules can achieve high efficiency up to 88%. The module offers direct cooling of dissipative components for excellent thermal performance. Standard features include remote On/Off, remote sense, output voltage adjustment, over voltage, over current and over temperature protection. The CQB100W series also have the following options: remote On/Off (positive or negative).

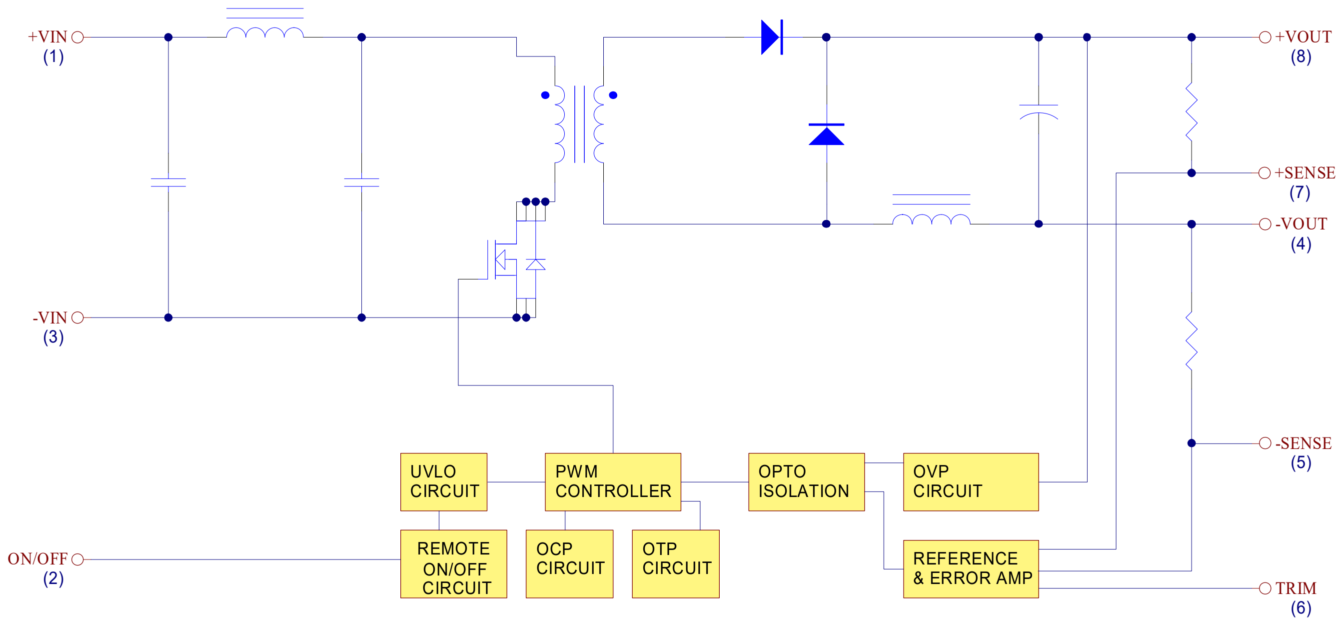
2. DC-DC Converter Features

- 100W Isolated Output
- Quarter-Brick Package
- 4:1 Input Range
- Efficiency to 88%
- Input Under Voltage Lockout Protection
- Regulated Output
- Over Temperature Protection
- Over Voltage/Current Protection
- Remote On/Off
- 1500VDC Isolation
- Safety Meets IEC/EN/UL 62368-1

3. Electrical Block Diagram



Electrical Block Diagram for 5Vout and 3.3Vout



Electrical Block Diagram for other modules

4. Technical Specifications

(All specifications are typical at nominal input, full load at 25℃ unless otherwise noted.)

4.1 Absolute Maximum Ratings

Stresses in excess of the absolute maximum ratings can cause permanent damage to the device. These are absolute stress ratings only, functional operation of the device is not implied at these or any other conditions in excess of those given in the operations sections of the data sheet. Exposure to absolute maximum ratings for extended periods can adversely affect the device reliability.

Parameter	Device	Symbol	Min	Typ.	Max	Unit
Input Voltage						
Continuous	24SXX 48SXX	V_{IN}	-0.3 -0.3		36 75	V_{dc}
Transient (100ms)	24SXX 48SXX	$V_{IN,trans}$			50 100	V_{dc}
Operating case Temperature (Note 1)	All	T_{CASE}	-40		100	℃
Storage Temperature	All	T_{stg}	-40		105	℃
I/O Isolation Voltage	All	—	—	1500	—	V_{dc}
Note1: see de-rating curve for application						

4.2 Electrical Specifications

Input Specifications

Parameter	Device	Symbol	Min	Typ.	Max	Unit
Operating Input Voltage	24SXX 48SXX	V_{IN}	9.0 18.0	24.0 48.0	36.0 75.0	V_{dc}
Input Under-Voltage Lockout						
Turn-On Voltage Threshold	24SXX 48SXX	V_{IN} (on)	8.0 16.0	8.8 17.0	9.0 18.0	V_{dc}
Turn-Off Voltage Threshold	24SXX 48SXX	V_{IN} (off)	7.5 15.0	8.0 16.0	8.5 17.0	V_{dc}
Maximum Input Current						
100% Load, 9 Vin, trim up	24SXX	$I_{IN,max}$			14.2	A
100% Load, 18 Vin, trim up	48SXX	$I_{IN,max}$			10.8	A
Input Ripple Current, Peak-to-Peak(Note1)	24SXX 48SXX	$I_{IN,p-p}$			30 50	mA
Input No Load Current (Vin=Nominal Input Voltage)	24S3V3,05 24S12,15,24 48S3V3,05 48S12,15,24	$I_{IN,No Load}$		120 80 60 30		mA
Input Standby Current	All	$I_{IN,standby}$			10	mA
Inrush Transient	24SXX 48SXX	I^2t			1.0 0.5	A^2s
Recommended Input Fuse (Note 2)	24SXX 48SXX				20 15	A
Input Capacitance (External) (Note 3)	24SXX 48SXX			100 47		μF
Notes:	Note1: 5Hz to 20MHz, 12μH source impedance; V_{IN}=$V_{IN,min}$ to $V_{IN,max}$, I_o= I_{omax} ; test setup see Fig.1. Note2: Slow Blow/Antisurge HRC recommended 200V Rating. Note3: Recommended customer added capacitance, <0.7Ω ESR.					

Output Specifications

Parameter	Device	Symbol	Min	Typ.	Max	Units
Output Voltage Set-point	3.3V_{dc}	$V_{o,set}$	3.25	3.3	3.35	V_{dc}
	5V_{dc}	$V_{o,set}$	4.92	5	5.08	
	12V_{dc}	$V_{o,set}$	11.82	12	12.18	
	15V_{dc}	$V_{o,set}$	14.77	15	15.23	
	24V_{dc}	$V_{o,set}$	23.64	24	24.36	
Output Voltage Regulation:						
Line (low line to high line)	All		—	—	0.2	% $V_{o,\ nom}$
Load (Io=Io,min to Io,max)	All		—	—	0.2	
Temperature Coefficient	All				±0.03	%/℃
Output Ripple and Noise (Note1)						
Peak-to-Peak (5Hz to 20MHz bandwidth)	3.3~5V				100	mV
	12~15V				150	
	24V				240	
RMS (5Hz to 20MHz bandwidth)	3.3~5V				40	mV
	12~15V				60	
	24V				100	
Output Voltage Trim Range	All	$V_{o,trim}$	-10	—	10	% $V_{o,\ nom}$
Remote-sense Compensation	All				10	% $V_{o,\ nom}$
Output Current	3.3 V_{dc}	I_o	0	—	30.0	A_{dc}
	5.0 V_{dc}	I_o	0	—	20.0	
	12 V_{dc}	I_o	0	—	8.30	
	15 V_{dc}	I_o	0	—	6.70	
	24 V_{dc}	I_o	0	—	4.17	
Output Current Limit Inception (V_o = 90% of V_o, set) Hiccup mode 110%-160%	3.3 V_{dc}	$I_{o,lim}$	33	40	48	A_{dc}
	5.0 V_{dc}	$I_{o,lim}$	22	27	32	
	12 V_{dc}	$I_{o,lim}$	9.13	11.2	13.28	
	15 V_{dc}	$I_{o,lim}$	7.37	9	10.72	
	24 V_{dc}	$I_{o,lim}$	4.59	5.62	6.67	
Efficiency (V_{IN} = Nominal Voltage; I_0 = $I_{0,max}$;T_A = 25 ℃)	24S3V3	η	—	86	—	%
	24S05	η	—	86.5	—	
	24S12	η	—	86.5	—	
	24S15	η	—	86.5	—	
	24S24	η	—	87	—	
	48S3V3	η	—	88	—	
	48S05	η	—	88	—	
	48S12	η	—	88	—	
	48S15	η	—	88	—	
	48S24	η	—	88	—	
Output Overvoltage (hiccup mode) 115-140%	3.3 V_{dc}	$V_{o,limit}$	3.79	3.96	4.6	V_{dc}
	5.0 V_{dc}	$V_{o,limit}$	5.75	6.0	7.0	
	12 V_{dc}	$V_{o,limit}$	13.8	14.4	16.8	
	15 V_{dc}	$V_{o,limit}$	17.2	18.0	21.0	
	24 V_{dc}	$V_{o,limit}$	27.6	28.8	33.6	

Parameter	Device	Symbol	Min	Typ.	Max	Unit
Over Temperature Shutdown (case)	All	$T_{CASE,ref}$	—	110	—	°C
Over Temperature Restart Hysteresis	All	$T_{CASE,hys}$	—	10	—	°C
Output Capacitance (External)	3.3~5V_{dc} 12~24V_{dc}	C_O	— —	— —	1,0000 2,200	μF
Note1: Measured with 10uF Tantalum and 1uF ceramic capacitors at the converter output pins(Over all operating input voltage, resistive load, and temperature conditions until end of life).						

Dynamic Specifications

Parameter	Device	Symbol	Min	Typ.	Max	Unit
Output Voltage during Load Current Transient (Note1)						
Peak Deviation	3.3 V_{dc} Others	V_{pk}			7 5	%V_o
Settling Time (<1% normal output)	All	t_s		200	500	μs
Turn-On Transient						
Turn-On Delay (Note2)	All	Td1		20	75	ms
Turn-On Delay (Note3)	All	Td2		110	250	ms
Output Rise time (Note4)	All	Tr		10	50	ms
Notes:	Note1: Load change from 50% to 75% to 50% of Io,max; $\Delta I_o/\Delta t$=0.1A/μs; $V_{IN}=V_{IN,nom}$; T_A=25°C ; 330μF Aluminum external capacitance and 1uF ceramic capacitor. Note2: Power applied first, then enable (time from enable applied until V_O=10% of $V_{o,set}$); Note3: Enable first, then power applied (time from input $V_{IN}=V_{IN,min}$ until V_O=10% of $V_{o,set}$); Note4: Time for V_o to rise from 10% of V_o, set to 90% of $V_{o,set}$;					

4.3 Isolation Specifications

Parameter	Device	Symbol	Min	Typ.	Max	Unit
Isolation Capacitance	All	Ciso	—	1000	—	pF
Isolation Resistance	All	Riso	10	—	—	MΩ

4.4 Feature Specifications

Parameter	Device	Symbol	Min	Typ.	Max	Units
Negative Remote Control (On/Off)						
Logic high-Unit off(Typ=Open Collector)	All		3.5		75.0	V_{dc}
Logic low-Unit On	All		-0.1		0.8	V_{dc}
Sink current	All				1	mA
Source current	All				1	mA
Positive Remote Control (On/Off)						
Logic high-Unit off	All		-0.1		0.8	V_{dc}
Logic low-Unit On(Typ=Open Collector)	All		3.5		75.0	V_{dc}
Sink current	All				1	mA
Source current	All				1	mA
Switching Frequency	24SXX 48SXX	f_{sw}		220 250		kHz
Mean Time Between Failure (Note1)	All	MTBF		600,000		Hours
Weight	All			66		g
Note1: Calculated according to MIL-HDBK-217F, GB 25℃ Full Load						

5. Main Features and Functions

5.1 Operating Temperature Range

The CQB100W series converters can be operated within a wide case temperature range of -40℃ to 100℃. Consideration must be given to the derating curves when ascertaining maximum power that can be drawn from the converter. The maximum power drawn from Quarter brick models is influenced by usual factors, such as:

- Input voltage range
- Output load current
- Forced air or natural convection

5.2 Output Voltage Adjustment

Section 6.8 describes in detail how to trim the output voltage with respect to its set point. The output voltage on all models is adjustable within the range of +10% to -10%.

5.3 Over Current Protection

The converter is protected against over current or short circuit conditions. At the instance of current-limit inception, the module enters a hiccup mode of operation, whereby it shuts down and automatically attempts to restart. While the fault condition exists, the module will remain in this hiccup mode, and can remain in this mode until the fault is cleared. The unit operates normally once the output current is reduced back into its specified range.

5.4 Output Overvoltage Protection

The converter is protected against output over voltage conditions. When the output voltage is higher than the specified range, the module enters a hiccup mode of operation. The operation is identical with over current protection.

5.6 Remote On/Off

The ON/OFF input Pin permits the user to turn the power module on or off via a system signal. Two remote on/off options are available. Positive logic turns the module on during a logic high voltage on the ON/OFF pin, and off during a logic low. Negative logic remote On/Off, turns the module off during a logic high and on during a logic low. The ON/OFF pin is internally pulled up through a resistor. A properly debounced mechanical switch, open collector transistor, or FET can be used to drive the input of the ON/OFF pin.
If not using the remote on/off feature:
For positive logic, leave the ON/OFF pin open.
For negative logic, short the ON/OFF pin to VIN(-).

5.7 UVLO (Undervoltage Lock Out)

Input undervoltage lockout is standard with this converter. At input voltages below the input undervoltage lockout limit, the module operation is disabled.

5.8 Over Temperature Protection

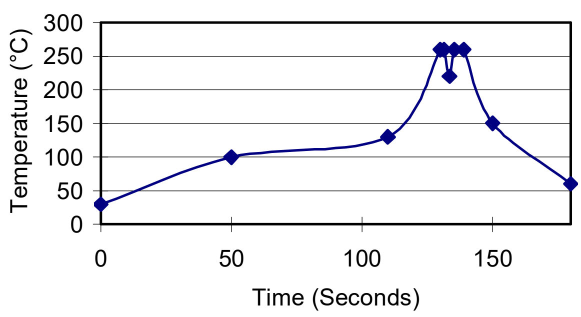
These modules have an over temperature protection circuit to safeguard against thermal damage. When the case temperature rises above over temperature shutdown threshold, the converter will shut down to protect itself from overheating. The module will automatically restarts after it cools down.

6. Applications

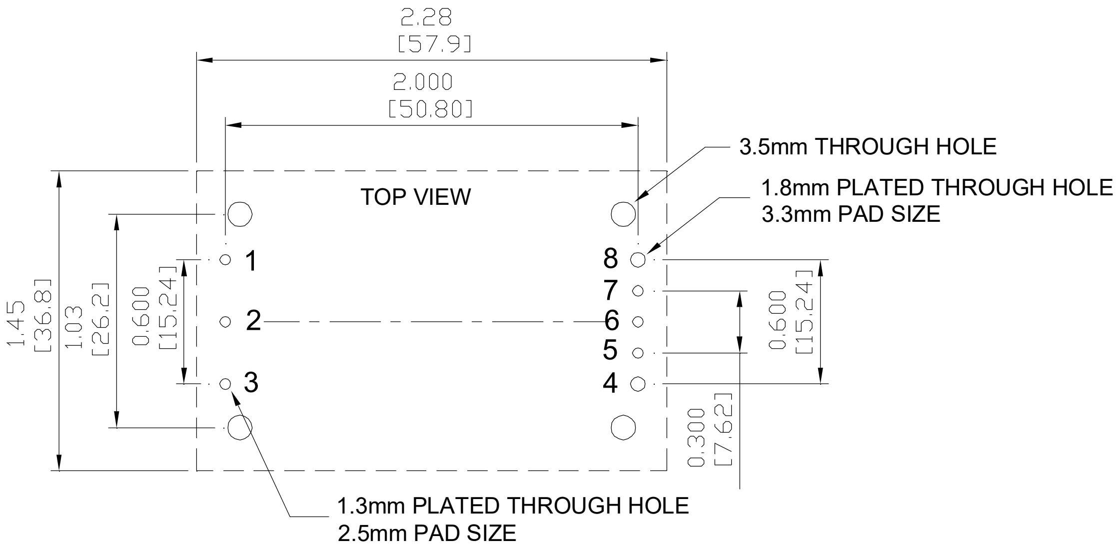
6.1 Recommended Layout, PCB Footprint and Soldering Information

The system designer or end user must ensure that metal and other components in the vicinity of the converter meet the spacing requirements for which the system is approved. Low resistance and inductance PCB layout traces are the norm and should be used where possible. Due consideration must also be given to proper low impedance tracks between power module, input and output grounds. The recommended soldering profile and PCB layout are shown below.

Lead Free Wave Soldering Profile



Note:
1. Soldering Materials : Sn/Cu/Ni
2. Ramp up rate during preheat: 1.4 °C/Sec (From 50°C to 100°C)
3. Soaking temperature: 0.5 ℃/Sec (From 100°C to 130°C), 60±20 seconds
4. Peak temperature: 260°C, above 250°C 3~6 Seconds
5. Ramp up rate during cooling: -10.0 °C/Sec (From 260°C to 150°C)



6.2 Convection Requirements for Cooling

To predict the approximate cooling needed for the Quarter brick module, refer to the power derating curves in section 6.4. These derating curves are approximations of the ambient temperatures and airflows required to keep the power module temperature below its maximum rating. Once the module is assembled in the actual system, the module's temperature should be monitored to ensure it does not exceed 100℃ as measured at the center of the top of the case (thus verifying proper cooling).

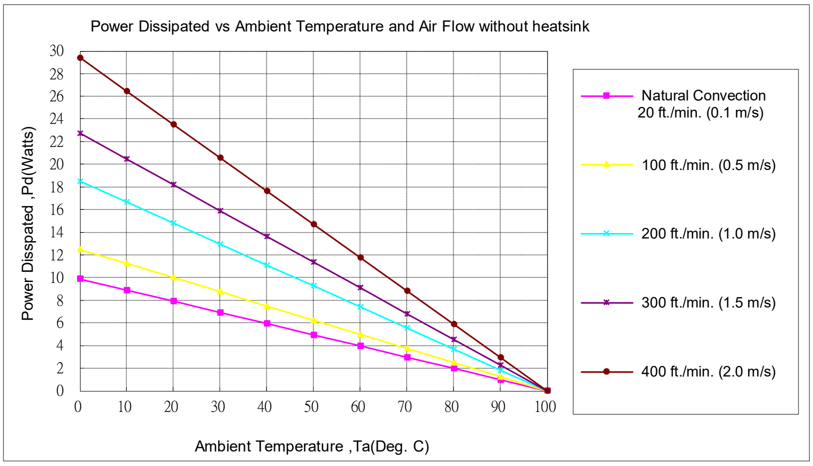
6.3 Thermal Considerations

The power module operates in a variety of thermal environments; however, sufficient cooling should be provided to help ensure reliable operation of the unit. Heat is removed by conduction, convection, and radiation to the surrounding environment. The power output of the module should not be allowed to exceed rated power (V_{o_set} x I_{o_max}).
The power modules have through-threaded,M3 x0.5 mounting holes, which enable heat sinks or cold plates to be attached to the module. Thermal de-rating with heat sinks is expressed by using the overall thermal resistance of the module(R_{ca}).

6.4 Power Derating

The operating case temperature range of CQB100W series is -40℃ to +100℃. When operating the CQB100W series, proper de-rating or cooling is needed. The maximum case temperature under any operating condition should not exceed 100℃.

Forced Convection Power De-rating without Heat Sink



AIR FLOW RATE	TYPICAL R_{ca}
Natural Convection 20ft./min. (0.1m/s)	10.1 ℃/W
100 ft./min. (0.5m/s)	8.0 ℃/W
200 ft./min. (1.0m/s)	5.4 ℃/W
300 ft./min. (1.5m/s)	4.4 ℃/W
400 ft./min. (2.0m/s)	3.4 ℃/W

Example (without heatsink):
What is the minimum airflow necessary for a CQB100W-48S05 operating at nominal line voltage, an output current of 20A, and a maximum ambient temperature of 40℃?

Solution:

Given: Vin=48V_{dc}, Vo=5V_{dc}, Io=20A

Determine Power dissipation (P_d):

$P_d=P_i-P_o=P_o(1-\eta)/\eta$

$P_d=5.0\times20\times(1-0.88)/0.88=13.64$Watts

Determine airflow:

Given: P_d=13.64W and T_a=40℃

Check above Power de-rating curve:

Airflow ≦ 400 ft./min.

Verifying:

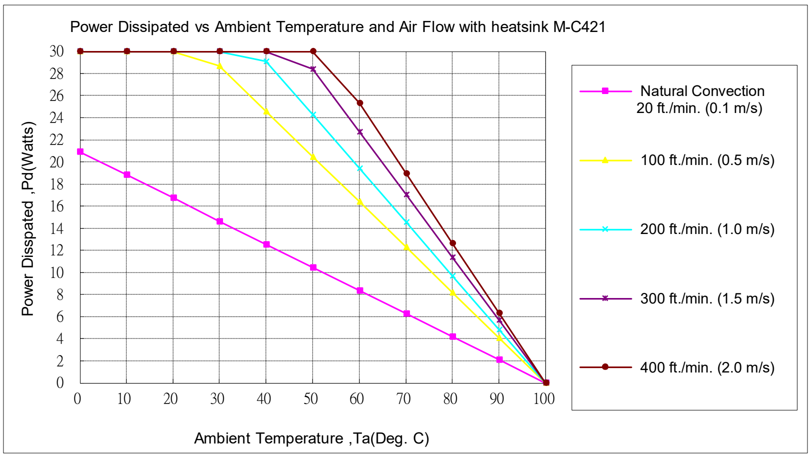
The maximum temperature rise $\triangle T = P_d \times R_{ca}=13.64\times3.4=46.4$℃

The maximum case temperature $T_c=T_a+\triangle T=86.4$℃ <100℃

Where:

The Rca is thermal resistance from case to ambience.

The Ta is ambient temperature and the T_c is case temperature



AIR FLOW RATE	TYPICAL R_{ca}
Natural Convection 20ft./min. (0.1m/s)	4.78 ℃/W
100 ft./min. (0.5m/s)	2.44 ℃/W
200 ft./min. (1.0m/s)	2.06 ℃/W
300 ft./min. (1.5m/s)	1.76 ℃/W
400 ft./min. (2.0m/s)	1.58 ℃/W

Example with heatsink QBT210 (M-C421):
What is the minimum airflow necessary for a CQB100W-48S05 operating at nominal line voltage, an output current of 20A, and a maximum ambient temperature of 40℃?

Solution:

Given: Vin=48V_{dc}, Vo=5V_{dc}, Io=20A

Determine Power dissipation (P_d):

Pd=Pi-Po=Po(1-η)/η
Pd=5.0×20×(1-0.88)/0.88=13.64Watts

Determine airflow:

Given: P_d=13.64W and T_a=40℃

Check above Power de-rating curve:

Airflow ≦ 100 ft./min

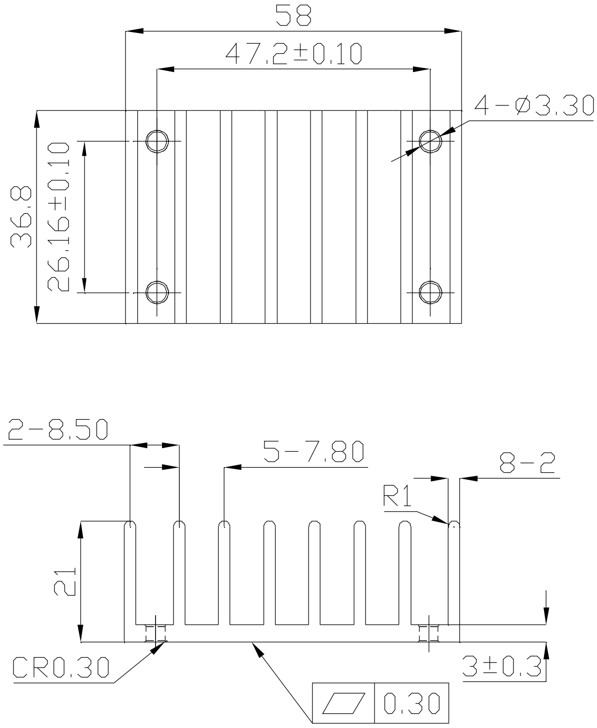
Verifying:

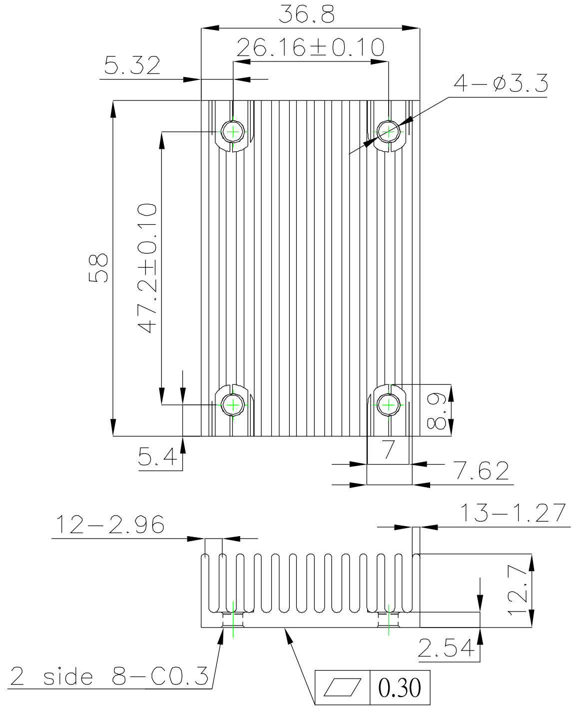
The maximum temperature rise △T = P_d × R_{ca}=13.64×2.44=33.28℃
The maximum case temperature Tc=Ta+△T=73.28℃ <100℃

Where:

The R_{ca} is thermal resistance from case to ambience.
The Ta is ambient temperature and the Tc is case temperature

6.5 Quarter Brick Heat Sinks:





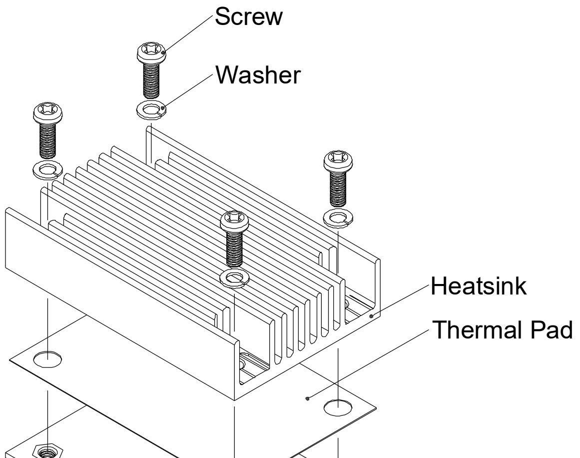
All Dimensions in mm

QBT210 (M-C421) G6620510201
Transverse Heat Sink

Rca: 4.78 ℃/W (typ.), At natural convection
2.44 ℃/W (typ.), At 100LFM
2.06 ℃/W (typ.), At 200LFM
1.76 ℃/W (typ.), At 300LFM
1.58 ℃/W (typ.), At 400LFM

QBL127 (M-C448) G6620570202
Longitudinal Heat Sink

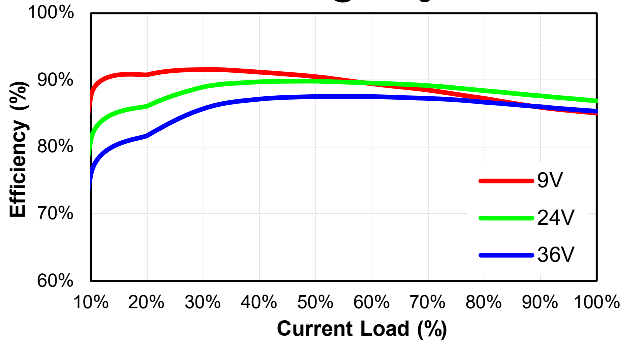
Rca: 5.61 ℃/W (typ.), At natural convection
4.01 ℃/W (typ.), At 100LFM
3.39 ℃/W (typ.), At 200LFM
2.86 ℃/W (typ.), At 300LFM
2.49 ℃/W (typ.), At 400LFM



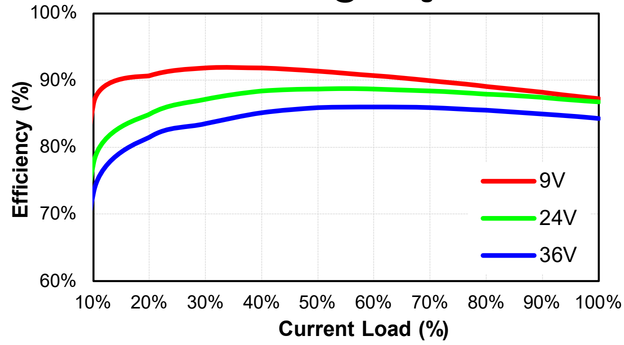
Heatsink：QBL127 (M-C448)
QBT210 (M-C421)
Thermail Pad PQ01：SZ35.8x56.9x0.25mm
Screw & Washer K308W：SMP+WS M3x8L
Recommended torque 4-8 Kgf-cm

6.6 Efficiency VS. Load

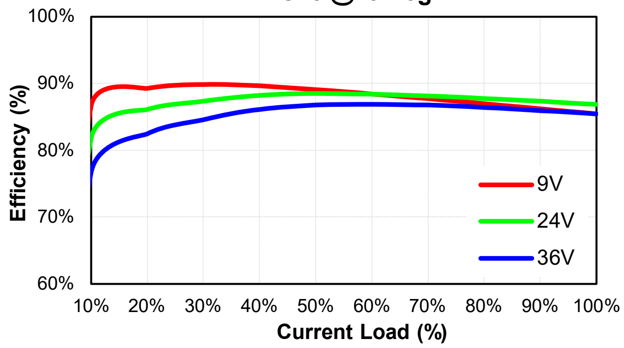
CQB100W-24S3V3
Eff Vs Io @25 Deg. C



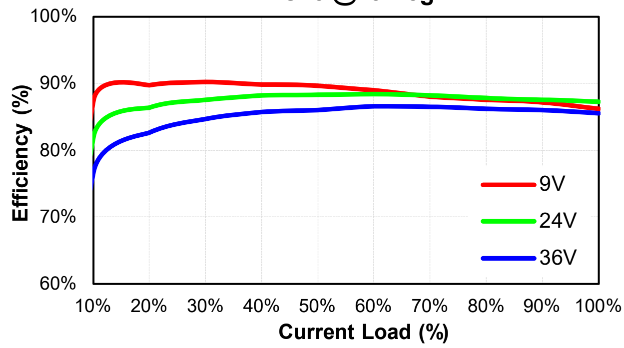
CQB100W-24S05
Eff Vs Io @25 Deg. C



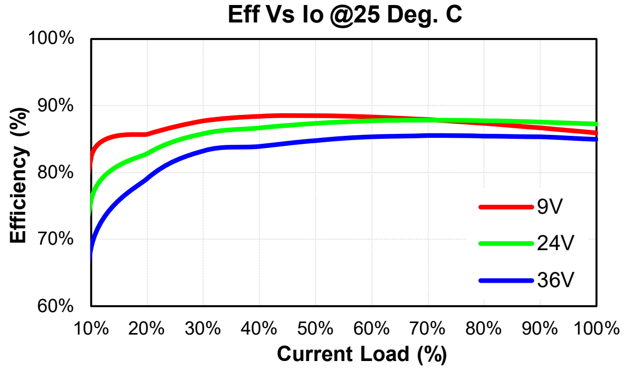
CQB100W-24S12
Eff Vs Io @25 Deg. C



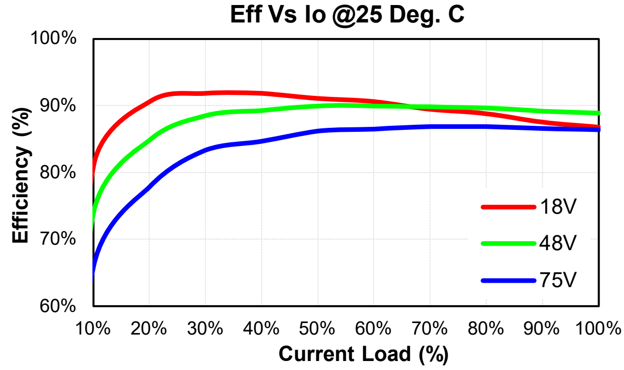
CQB100W-24S15
Eff Vs Io @25 Deg. C

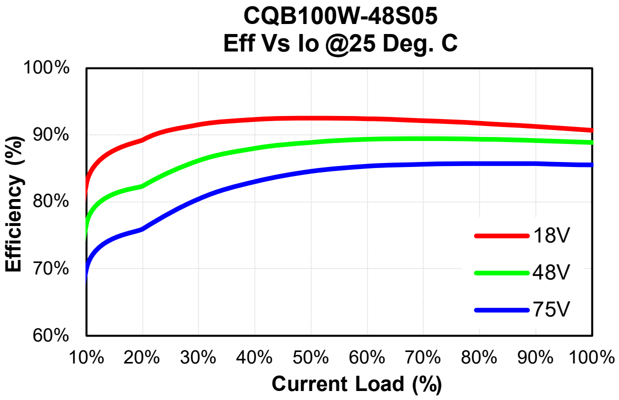


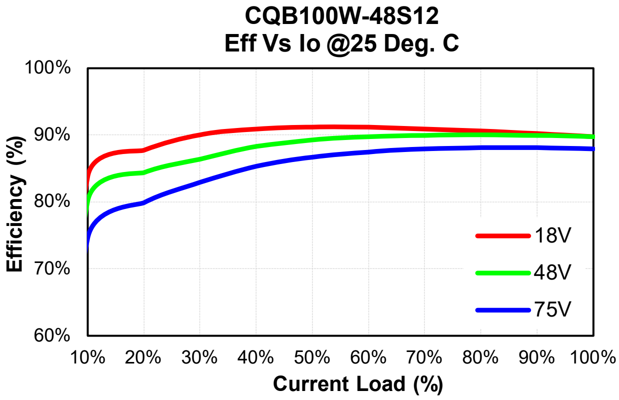
CQB100W-24S24
Eff Vs Io @25 Deg. C

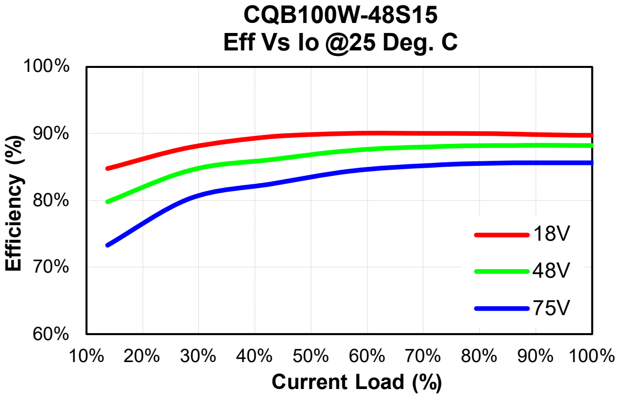


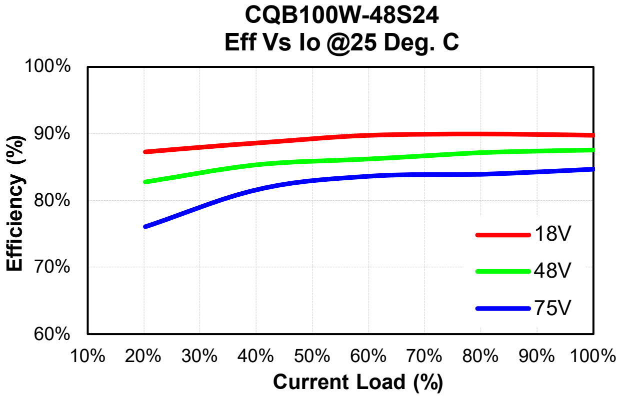
CQB100W-48S3V3
Eff Vs Io @25 Deg. C




CQB100W-48S05
Eff Vs Io @25 Deg. C
100%
90%
80%
70%
60%
Efficiency (%)
18V
48V
75V
10% 20% 30% 40% 50% 60% 70% 80% 90% 100%
Current Load (%)


CQB100W-48S12
Eff Vs Io @25 Deg. C
100%
90%
80%
70%
60%
Efficiency (%)
18V
48V
75V
10% 20% 30% 40% 50% 60% 70% 80% 90% 100%
Current Load (%)


CQB100W-48S15
Eff Vs Io @25 Deg. C
100%
90%
80%
70%
60%
Efficiency (%)
18V
48V
75V
10% 20% 30% 40% 50% 60% 70% 80% 90% 100%
Current Load (%)


CQB100W-48S24
Eff Vs Io @25 Deg. C
100%
90%
80%
70%
60%
Efficiency (%)
18V
48V
75V
10% 20% 30% 40% 50% 60% 70% 80% 90% 100%
Current Load (%)

6.7 Test Set-Up

The basic test set-up to measure parameters such as efficiency and load regulation is shown below. When testing the modules under any transient conditions please ensure that the transient response of the source is sufficient to power the equipment under test. We can calculate:

- Efficiency
- Load regulation and line regulation.

The value of efficiency is defined as:

$$\eta = \frac{Vo \times Io}{Vin \times Iin} \times 100\%$$

Where:
V_o is output voltage,
I_o is output current,
V_{in} is input voltage,
I_{in} is input current.

The value of load regulation is defined as:

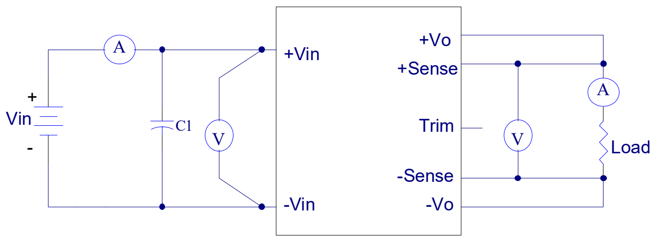
$$Load.reg = \frac{V_{FL} - V_{NL}}{V_{NL}} \times 100\%$$

Where:
V_{FL} is the output voltage at full load
V_{NL} is the output voltage at no load

The value of line regulation is defined as:

$$Line.reg = \frac{V_{HL} - V_{LL}}{V_{LL}} \times 100\%$$

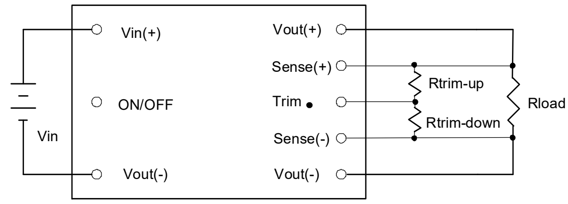
Where:
V_{HL} is the output voltage of maximum input voltage at full load. V_{LL} is the output voltage of minimum input voltage at full load.



CQB100W Series Test Setup

6.8 Output Voltage Adjustment

The Trim input permits the user to adjust the output voltage up or down 10%. This is accomplished by connecting an external resistor between the Trim pin and either the VO(+) pin or the VO(-) pin (COM pin), see Figure



Output voltage trim circuit configuration

The Trim pin should be left open if trimming is not being used. Connecting an external resistor ($R_{trim-down}$) between the Trim pin and the $V_{out(-)}$ (or Sense(-)) pin decreases the output voltage. The following equation determines the required external resistor value to obtain a down percentage output voltage change of Δ%

$$R_{trim-down} = \left[\frac{511}{\Delta\%} - 10.22 \right] k\Omega$$

Where

$$\Delta\% = \left(\frac{V_{o,set} - V_{desired}}{V_{o,set}} \right) \times 100$$

For example, to trim-down the output voltage of 12V module(CQB100W-48S12) by 5% to 11.4V, $R_{trim-down}$ is calculated as follow:

Δ%=5

$$R_{trim-down} = (\frac{511}{5} - 10.22) k\Omega$$

$$R_{trim-down} = 91.98 k\Omega$$

Connecting an external resistor (Rtrim-up) between the Trim pin and the V_{out} (+) (or Sense (+)) pin increases the output voltage. The following equations determine the required external resistor value to obtain a up percentage output voltage change of Δ%.

$$R_{trim-up} = \left[\frac{5.11 V_{out}(100 + \Delta\%)}{1.225 \times \Delta\%} - \frac{511}{\Delta\%} - 10.22 \right] k\Omega$$

Where

$$V_{out} = V_{o,set},\ \Delta\% = \left(\frac{V_{desired} - V_{o,set}}{V_{o,set}} \right) \times 100$$

For example, to trim-up the output voltage of 12V module(CQB100W-48S12) by 5% to 12.6V, $R_{trim-up}$ is calculated as follow:

Δ%=5

$$R_{trim-up} = (\frac{5.1 \times 12 \times (100 + 5)}{1.225 \times 5} - \frac{511}{5} - 10.22) k\Omega$$

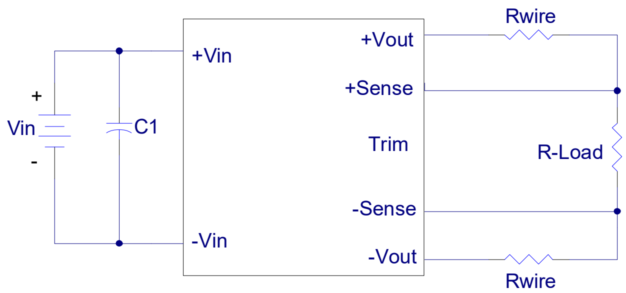
$$R_{trim-up} = 936.74 k\Omega$$

6.9 Output Remote Sensing

The CQB100W SERIES converter has the capability to remotely sense both lines of its output. This feature moves the effective output voltage regulation point from the output of the unit to the point of connection of the remote sense pins. This feature automatically adjusts the real output voltage of the CQB100W SERIES in order to compensate for voltage drops in distribution and maintain a regulated voltage at the point of load. The remote-sense voltage range is :

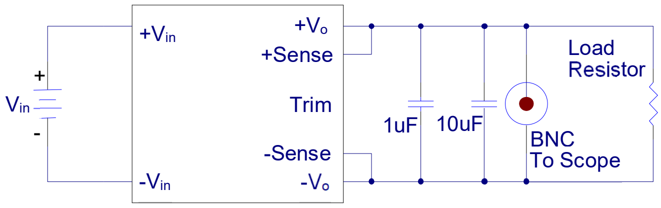
$[(+V_{out}) - (-V_{out})] – [(+Sense) – (-Sense)] \leqq 10\%$ of $V_{o\_nominal}$

If the remote sense feature is not to be used, the sense pins should be connected locally. The +Sense pin should be connected to the +Vout pin at the module and the -Sense pin should be connected to the -Vout pin at the module. This is shown in the schematic below.



Note:
Although the output voltage can be varied (increased or decreased) by both remote sense and trim, the maximum variation for the output voltage is the larger of the two values not the sum of the values. The output power delivered by the module is defined as the voltage at the output terminals multiplied by the output current. Using remote sense and trim can cause the output voltage to increase and consequently increase the power output of the module if output current remains unchanged. Always ensure that the output power of the module remains at or below the maximum rated power. Also be aware that if $V_{o.set}$ is below nominal value, $P_{out.max}$ will also decrease accordingly because $I_{o.max}$ is an absolute limit. Thus, $P_{out.max} = V_{o.set} \times I_{o.max}$ is also an absolute limit.

6.10 Output Ripple and Noise



Output ripple and noise is measured with 1.0uF ceramic and 10uF solid tantalum capacitors across the output.

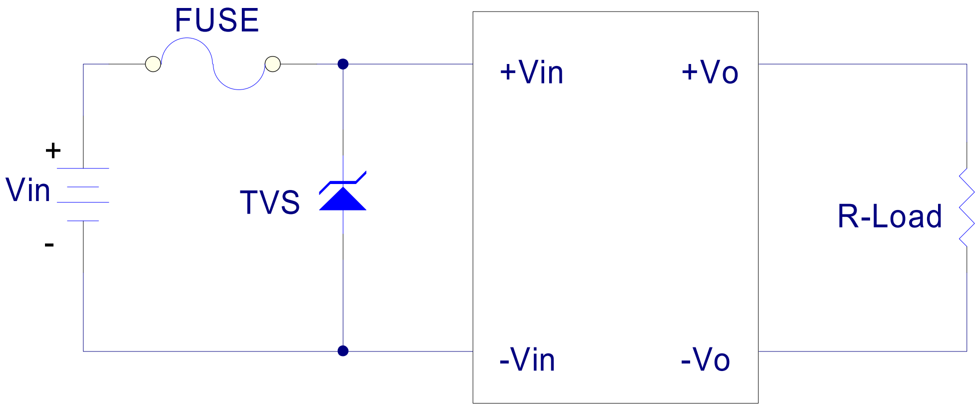
6.11 Output Capacitance

The CQB100W series converters provide unconditional stability with or without external capacitors. For good transient response, low ESR output capacitors should be located close to the point of load. PCB design emphasizes low resistance and inductance tracks in consideration of high current applications. Output capacitors with their associated ESR values have an impact on loop stability and bandwidth. These series converters are designed to work with load capacitance to see technical specifications.

7. Safety & EMC

7.1 Input Fusing and Safety Considerations

The CQB100W series converters have no internal fuse. In order to achieve maximum safety and system protection, always use an input line fuse. We recommended a 15A time delay fuse for the models. It is recommended that the circuit have a transient voltage suppressor diode (TVS) across the input terminal to protect the unit against surge or spike voltage and input reverse voltage (as shown).



7.2 EMC Considerations

EMI Test standard: EN55032 Class A Conducted Emission
Test Condition: Input Voltage: Nominal, Output Load: Full Load

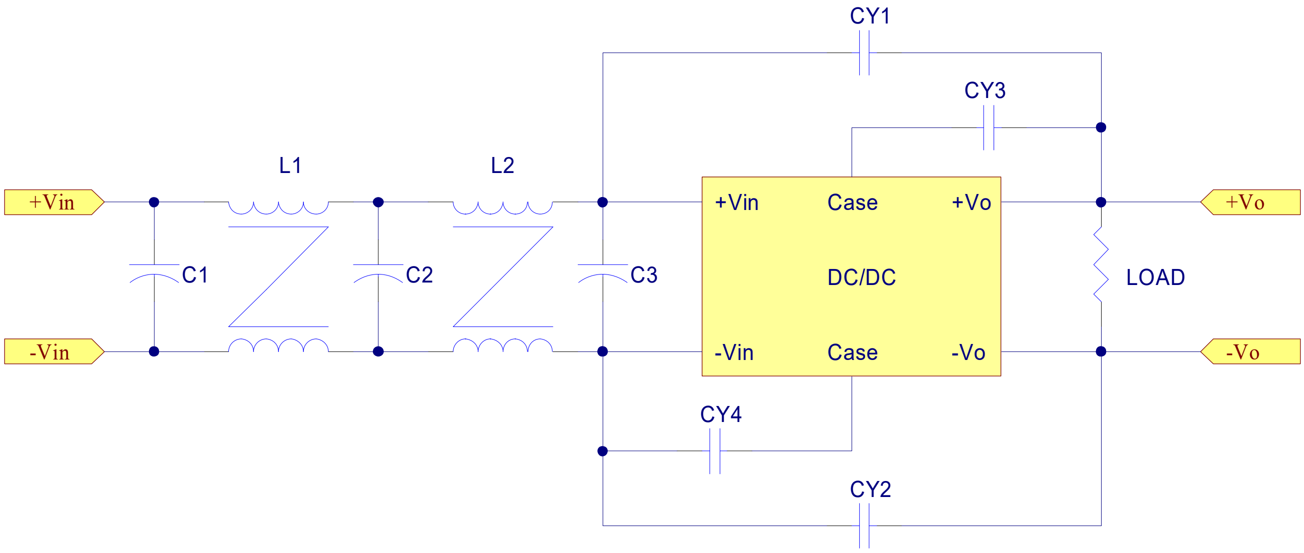


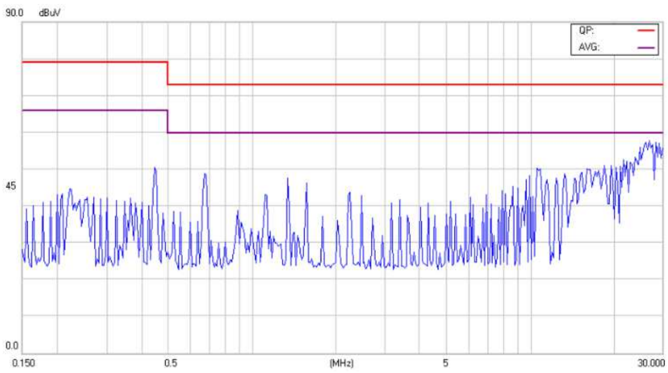
Figure1 Connection circuit for conducted EMI testing

CQB100W Series
Application Note V13

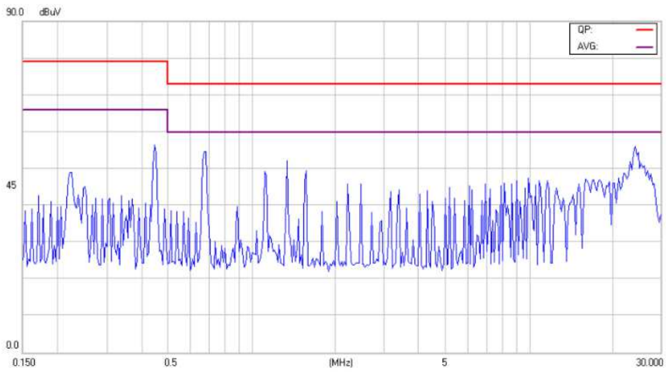
Model No.	C1	C2	C3	CY1, CY2	CY3, CY4	L1	L2
CQB100W-24S3V3	470uF/50V	470uF/50V	NC	4700pF	NC	5.0mH	short
CQB100W-24S05	470uF/50V	NC	470uF/50V	1000pF	2200pF	5.0mH	5.0mH
CQB100W-24S12	470uF/50V	470uF/50V	NC	1500pF	NC	5.0mH	short
CQB100W-24S15	470uF/50V	470uF/50V	NC	1500pF	NC	5.0mH	short
CQB100W-24S24	470uF/50V	470uF/50V	NC	1500pF	NC	5.0mH	short
CQB100W-48S3V3	150uF/100V	150uF/100V	NC	NC	NC	0.5mH	short
CQB100W-48S05	150uF/100V	150uF/100V	NC	NC	NC	0.5mH	short
CQB100W-48S12	150uF/100V	150uF/100V	NC	NC	NC	0.5mH	short
CQB100W-48S15	150uF/100V	150uF/100V	NC	NC	NC	0.5mH	short
CQB100W-48S24	150uF/100V	150uF/100V	NC	NC	NC	0.5mH	short

Note:
C1, C2, C3: NIPPON CHEMI-CON KY series aluminum capacitor or equivalent.
CY1~CY4: TDK Y2 capacitor or equivalent.
L1, L2: CINCON P/N: G91CA826015 for 24Vin
VAKOS CORE: T16x9x5C, Wire: Φ0.9mm*1/11T for 48Vin

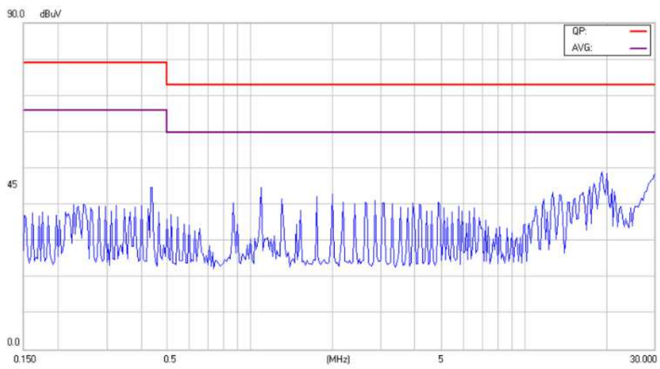
Conducted Class A of CQB100W-24S3V3



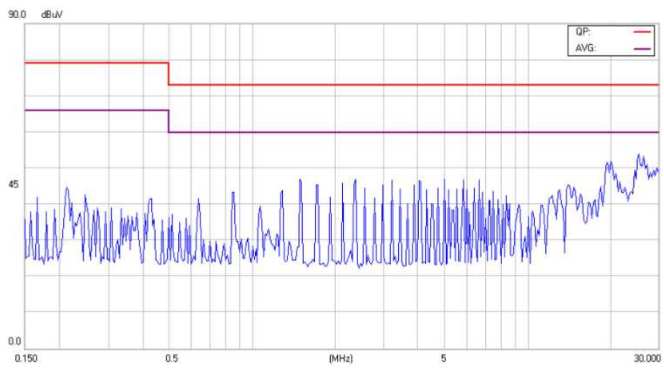
Conducted Class A of CQB100W-24S05



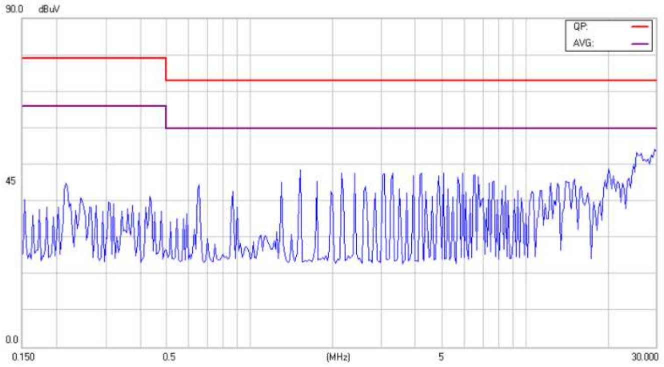
Conducted Class A of CQB100W-24S12



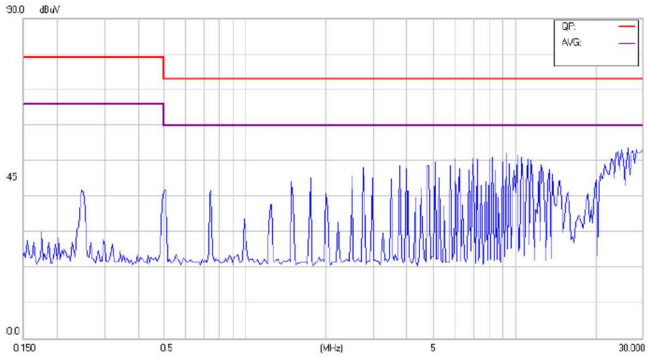
Conducted Class A of CQB100W-24S15



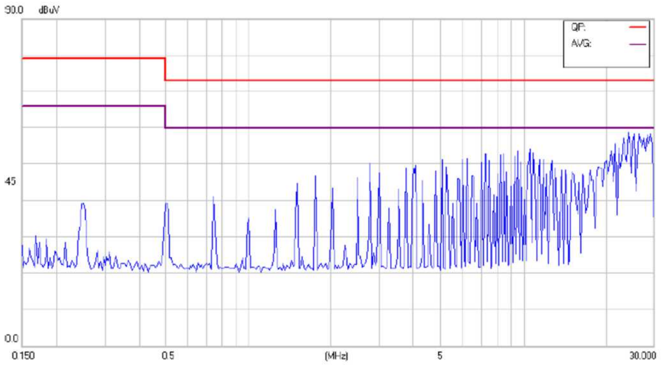
Conducted Class A of CQB100W-24S24



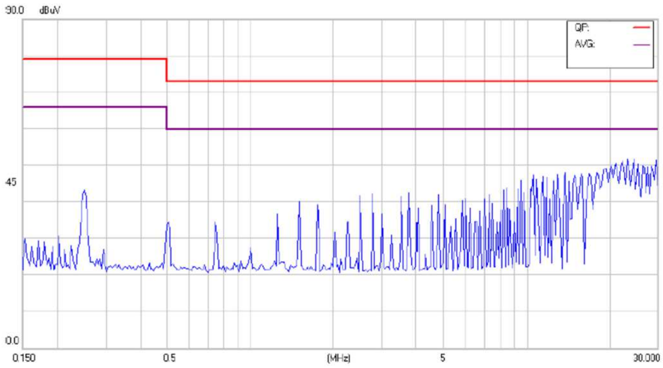
Conducted Class A of CQB100W-48S3V3



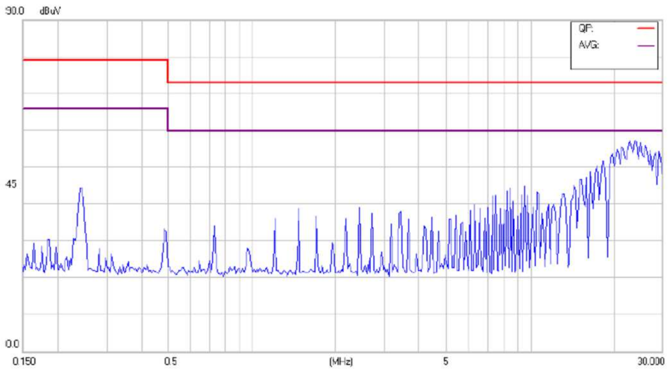
Conducted Class A CQB100W-48S05



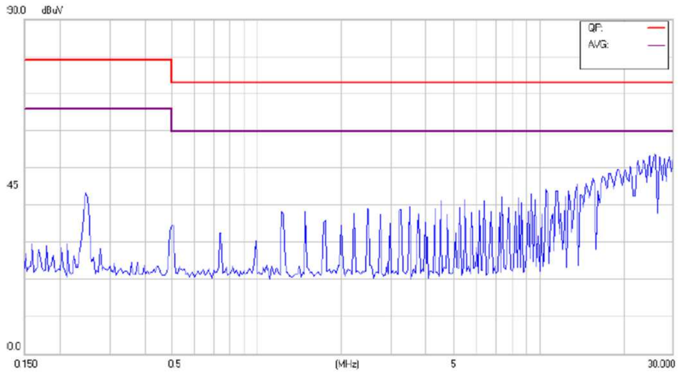
Conducted Class A of CQB100W-48S12



Conducted Class A CQB100W-48S15



Conducted Class A of CQB100W-48S24



8. Part Number

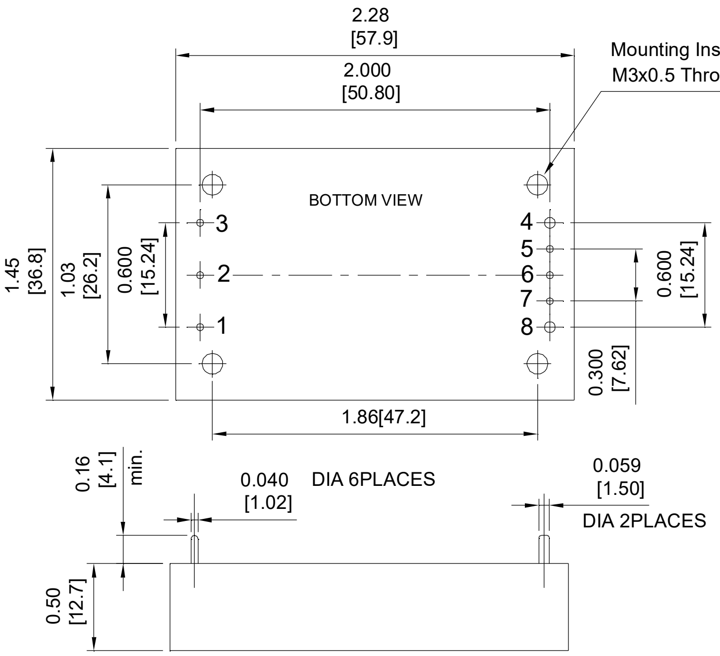
Format: CQB100W – II X OO LY

Parameter	Series	Nominal Input Voltage	Number of Outputs	Output Voltage	Remote ON/OFF Logic	Option	
Symbol	CQB100W	II	X	OO	L	Y	
Value	CQB100W	24:24Volts 48:48Volts	S: Single	3V3: 3.3 Volts 05: 05 Volts 12: 12 Volts 15: 15 Volts 24: 24 Volts	None: Positive N: Negative	C	Clear Mounting Insert

9. Mechanical Specifications

9.1 Mechanical Outline Diagrams

All Dimensions In Inches(mm)
Tolerances Inches: X.XX= ±0.02 , X.XXX= ±0.010
Millimeters: X.X= ±0.5 , X.XX=±0.25



PIN CONNECTION	
Pin	Function
1	+Vin
2	ON/OFF
3	-Vin
4	-Vout
5	-Sense
6	Trim
7	+Sense
8	+Vout

CQB100W Mechanical Outline Diagram

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